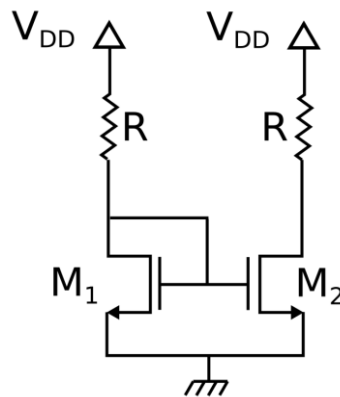


Unit 3: EXERCISES

Ex1. For the following circuit, compute the current of the transistors ($L = 1\mu\text{m}$).



Data: $V_{DD} = 5\text{V}$, $R = 100\text{ k}\Omega$, $\mu_n C_{ox} = 120\ \mu\text{A}/\text{V}^2$, $(W/L)_1 = (W/L)_2 = 5$, $V_{th} = 0.8\text{V}$.

Neglecting channel modulation and assuming saturation for both M_1 and M_2 :

$$I_{M1} = \frac{V_{DD} - V_{GS,M1}}{R}$$

$$I_{M1} = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{th})^2$$

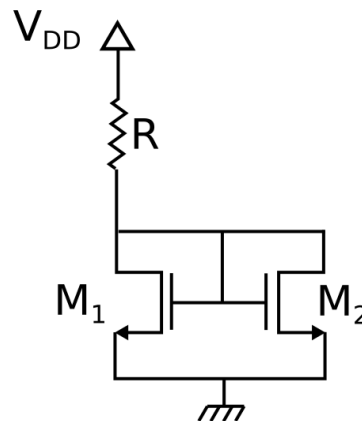
$$\rightarrow V_{GS,M1} = 1.16\text{V}, I_{M1} = 39\ \mu\text{A}$$

M_1 and M_2 have the same size ratio, so:

$$I_{M1} = I_{M2} = 39\ \mu\text{A}$$

In the file `ex1_current_mirror.asc` you can see a simulation of this example, showing similar results than the theoretical ones.

Ex2. Repeat Ex1 for the following circuit:



Data: $V_{DD} = 5V$, $R = 100 \text{ k}\Omega$, $\mu_n C_{ox} = 120 \mu\text{A}/\text{V}^2$, $(W/L)_1 = (W/L)_2 = 5$, $V_{th} = 0.8V$.

Neglecting channel modulation and assuming saturation for both M_1 and M_2 :

$$I_{M1} + I_{M2} = \frac{V_{DD} - V_{GS,M1}}{R}$$

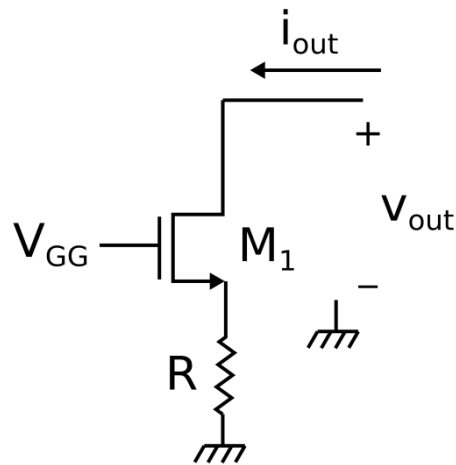
$$I_{M1} = I_{M2} = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{th})^2$$

$$\rightarrow V_{GS,M1} = 1.16V, I_{M1} = I_{M2} = 19.5 \mu\text{A}$$

Is it possible to replace M_1 and M_2 with one single device?

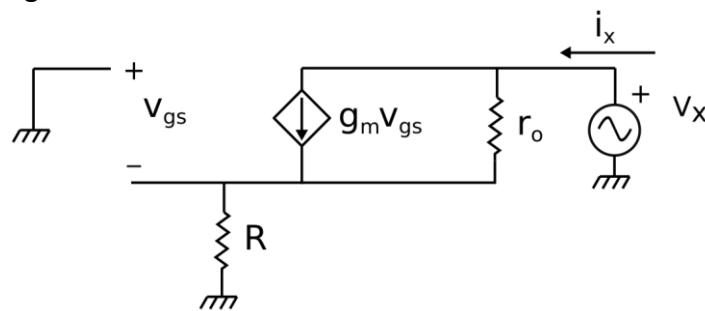
Yes, we can use one single device with twice the size ratio $\rightarrow (W/L) = 10$.

Ex3. Compute the output resistance of the following current source:



Data: $R = 50 \text{ k}\Omega$, $\mu_n C_{ox} = 110 \text{ }\mu\text{A}/\text{V}^2$, $(W/L)_1 = 2$, $i_{out,DC} = 10 \text{ }\mu\text{A}$, $\lambda = 0.04 \text{ V}^{-1}$

Using the small-signal model of M_1 :



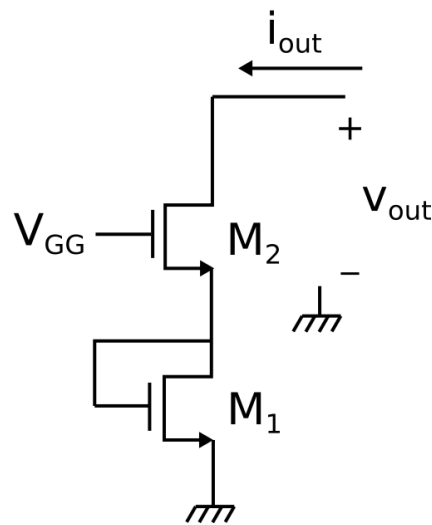
$$r_{out} = \frac{v_x}{i_x} = R + r_o + g_m r_o R$$

$$g_m = \sqrt{2\mu_n C_{ox} \frac{W}{L} I_D} = 66.33 \frac{\mu\text{A}}{\text{V}}$$

$$r_o = \frac{1}{\lambda I_D} = 2.5 \text{ M}\Omega$$

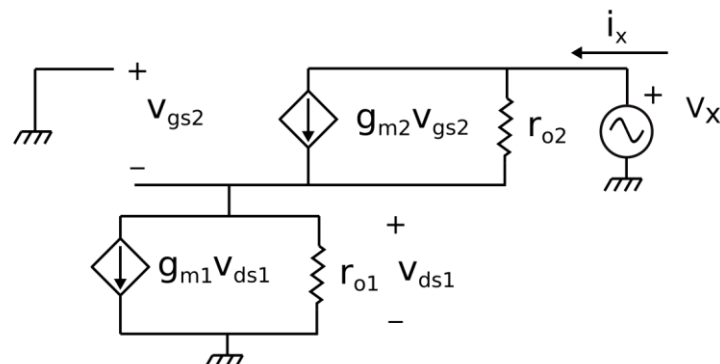
$$r_{out} = 10.84 \text{ M}\Omega$$

Ex4. Making use of the following circuit, compute $(W/L)_{M1}$ to have the same output resistance as in Ex3.



Data: $\mu_n C_{ox} = 110 \mu\text{A}/\text{V}^2$, $(W/L)_2=2$, $i_{\text{out,DC}} = 10 \mu\text{A}$, $\lambda=0.04 \text{ V}^{-1}$

Using the small-signal model:



$$r_{out} = \frac{v_x}{i_x} = \frac{r_{o1} r_{o2} (g_{m1} + g_{m2}) + r_{o1} + r_{o2}}{1 + g_{m1} r_{o1}}$$

$$g_{m2} = \sqrt{2\mu_n C_{ox} \frac{W}{L} I_D} = 66.33 \frac{\mu\text{A}}{\text{V}}$$

$$r_{o1} = r_{o2} = \frac{1}{\lambda I_D} = 2.5 \text{ M}\Omega$$

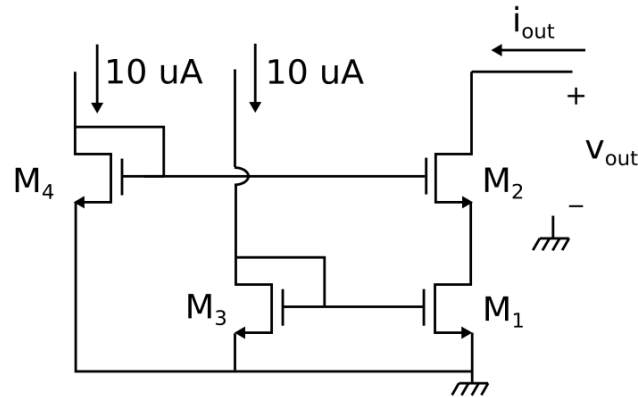
$$r_{out} = 11 \text{ M}\Omega$$

$$\rightarrow g_{m1} = \sqrt{2\mu_n C_{ox} \frac{W}{L} I_D} = 19.22 \frac{\mu\text{A}}{\text{V}}$$

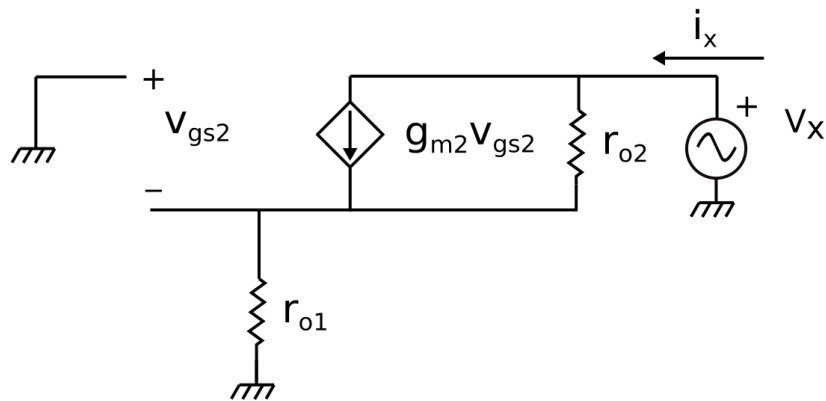
$$\left(\frac{W}{L}\right)_1 \approx 1/6$$

For instance, we can choose: $W = 1 \mu\text{m}$, $L = 6 \mu\text{m}$.

Ex5. Compute the output resistance of the following current source.



Data: $\mu_n C_{ox} = 110 \mu A/V^2$, $(W/L)_1=(W/L)_2=(W/L)_3=4$, $(W/L)_4=1$, $\lambda=0.04 V^{-1}$



Using the small-signal model:

$$r_{out} = \frac{v_x}{i_x} = r_{o1} + r_{o2} + g_{m1} r_{o1} r_{o2}$$

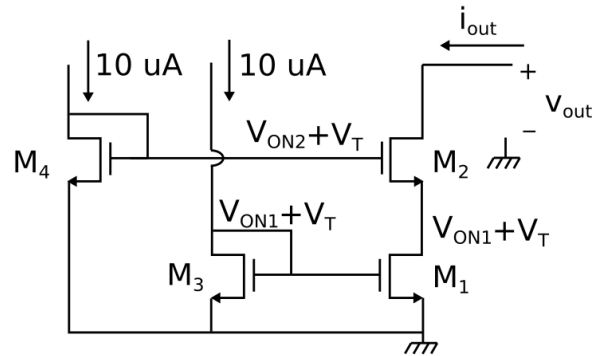
$$i_{out} = 10 \mu A$$

$$g_{m1} = \sqrt{2\mu_n C_{ox} \frac{W}{L} I_D} = 93.8 \frac{\mu A}{V}$$

$$r_{o1} = r_{o2} = \frac{1}{\lambda I_D} = 2.5 M\Omega$$

$$r_{out} = 591 M\Omega$$

What is the minimum V_{out} we may have?



Where V_{ON} is the overdrive voltage $\rightarrow V_{ON} = \sqrt{\frac{2I_D}{\mu_n C_{ox}(W/L)}}$

$$V_{D2} \geq V_{G2} - V_T$$

$$V_{out,min} = V_{ON2}$$

With $I_D = 10 \mu A$

$$\rightarrow V_{out,min} = 0.426 V$$



Exercises adapted from:

- Allen, P. E., & Holberg, D. R. (2011). CMOS analog circuit design. New York: Oxford University Press, USA.
- R Jacob Baker 2010 CMOS Circuit Design, Layout, and Simulation 3 rd ed Wiley IEEE Press